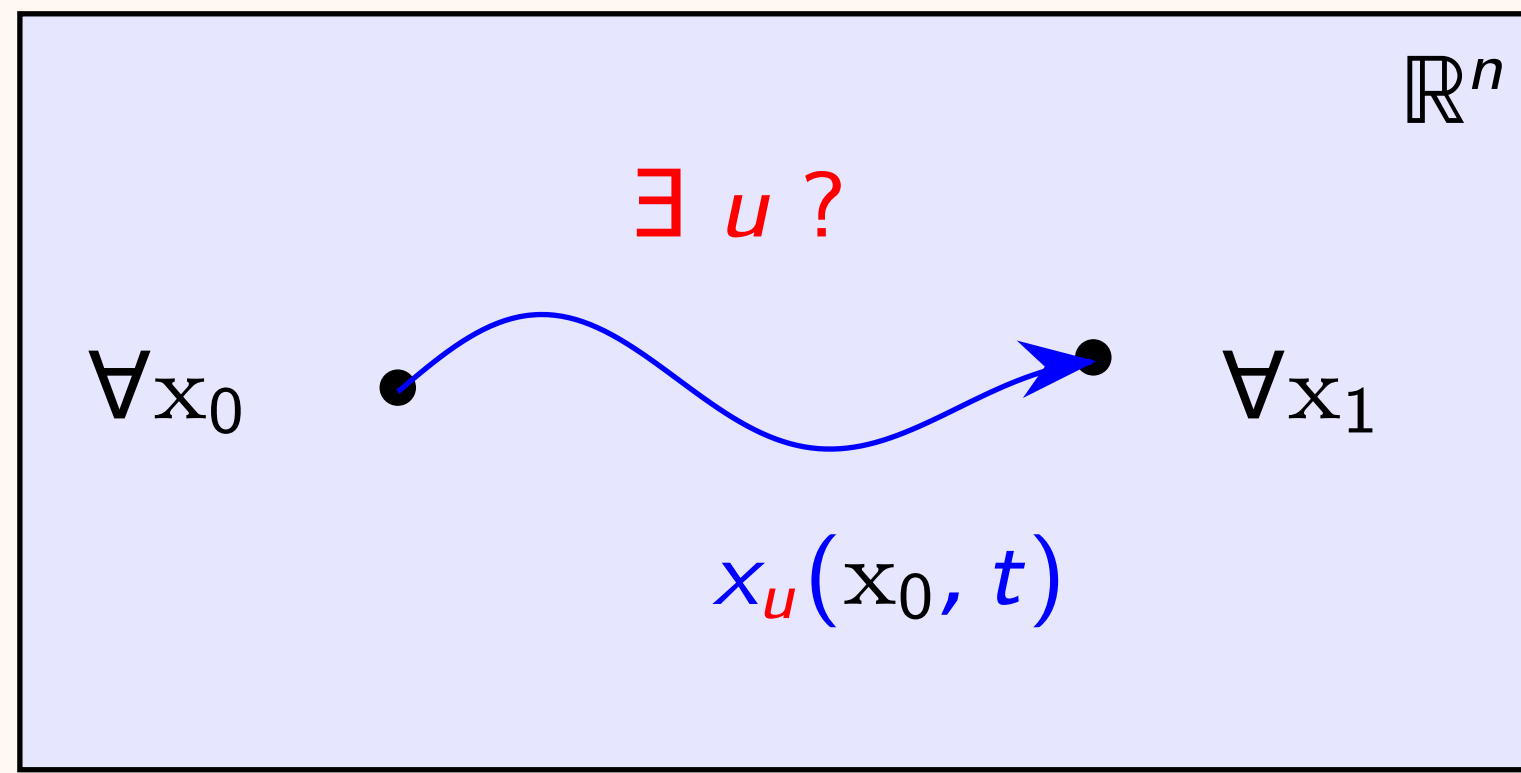


## Introduction

Consider for  $t \geq 0$  the Linear Time Invariant (LTI) system

$$\dot{x}(t) = Ax(t) + Bu(t).$$

- State controllability



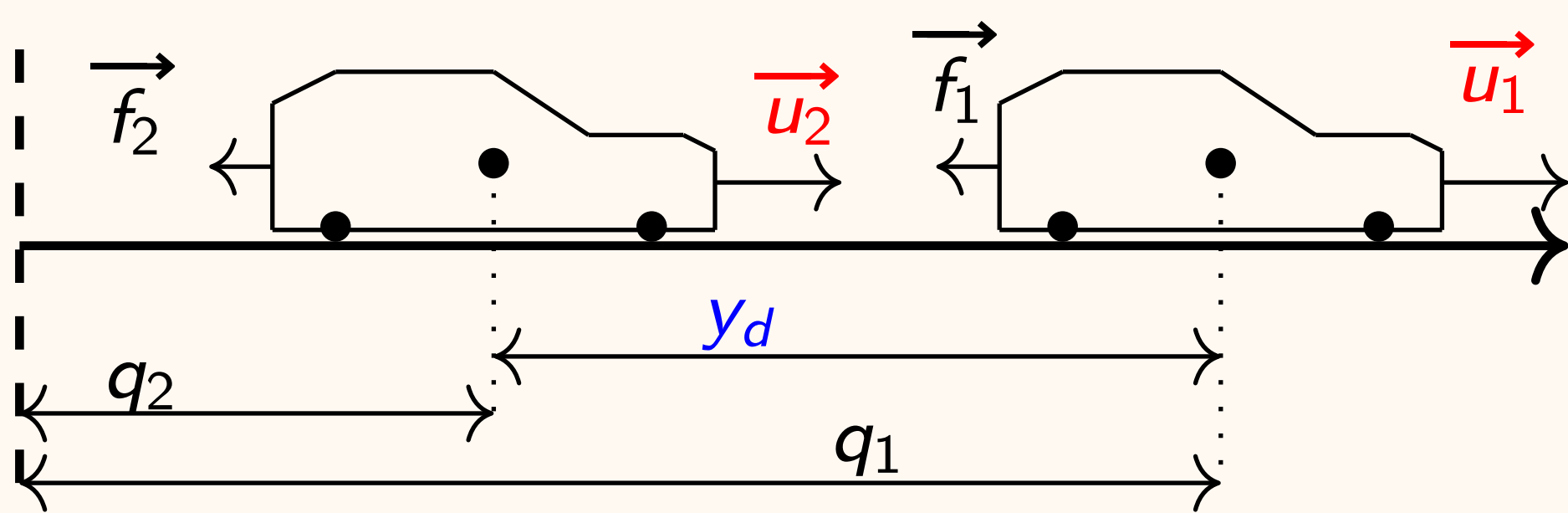
Topic very well understood from literature.

Particularity

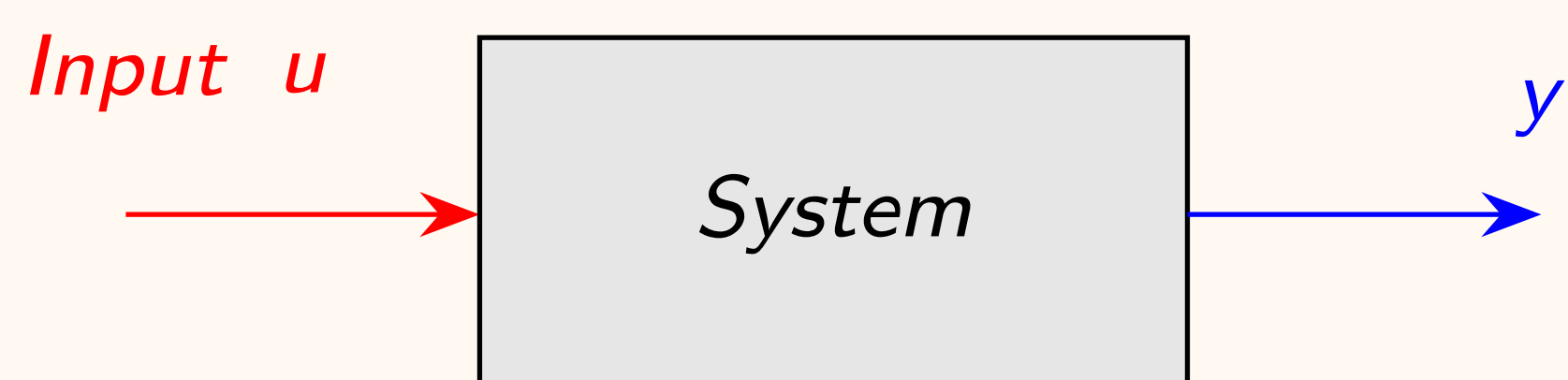
All the state variables are controlled.

## Motivation

- What if you do not will to control all the state variables ?



- More generally



Questions:

- Can we reach any benchmark output  $y_{ref}$ ?
- If YES, what is the suitable  $u$  to be used?

To answer these questions, Bertram and Sarachik introduced in [1] the notion of **Output controllability**.

## Framework: LTI systems

We assume that the output is given by

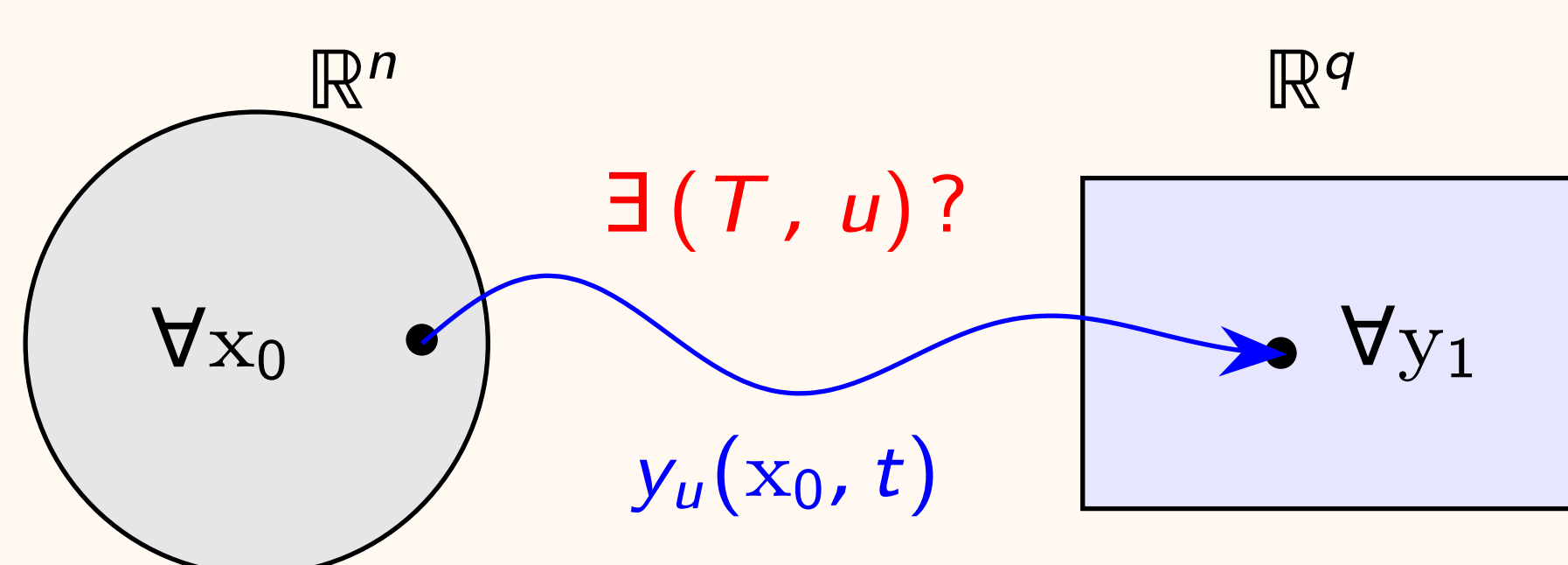
$$\begin{aligned} \dot{x}(t) &= Ax(t) + Bu(t), \\ y(t) &= Cx(t) + Du(t). \end{aligned} \quad (1)$$

$$x(t) \in \mathbb{R}^n, \quad u(t) \in \mathbb{R}^m, \quad y(t) \in \mathbb{R}^q.$$

Discussed notions:

State to output controllability.

## State to output controllability



$u$  continuous and  $y_u(x_0, T) = y_1$ .

Two other notions of Output controllability can be found in [2]. In these notions, we discuss how to steer an initial output value to a prescribed output.

## Theorem (Criteria)

The system (1) is state to output controllable if and only if one of the following conditions is fulfilled

- (i) There exists a time  $T > 0$  such that the linear map  $E_T^o : \mathcal{C}^0([0, T]; \mathbb{R}^m) \rightarrow \mathbb{R}^q$ , defined by

$$E_T^o(u) = \int_0^T Ce^{(T-\tau)A}Bu(\tau)d\tau + Du(T) \text{ is surjective.}$$

- (ii)  $\text{rk}[D \ CB \ CAB \ \dots \ CA^{n-1}B] = q$ . (Given in [3])

- (iii)  $\mathcal{K}_T := C \int_0^T e^{tA}BB^T e^{tA^T} dt C^T + DD^T > 0$  for some  $T > 0$ . (Also given in [3])

- (iv)  $\text{rk}(C|D) = q$  and  $\text{Im} \begin{pmatrix} C^T \\ D^T \end{pmatrix} \cap \left( \bigoplus_{\lambda \in \sigma(A)} E_\lambda \times \{0^m\} \right) = \{0\}$ , where

$E_\lambda = \ker(A_\lambda^T)^{n_\lambda} \cap \left( \bigcap_{k=0}^{n_\lambda-1} \ker B^T(A_\lambda^T)^k \right)$ ,  $A_\lambda = A - \lambda I_n$  and  $n_\lambda$ , the algebraic multiplicity of  $\lambda$  in the minimal polynomial of  $A$ .

- (v)  $\text{rk}(C|D) = q$  and  $\text{rk} \begin{pmatrix} K_{\lambda_1} & 0 & \dots & 0 & (C|D)^\perp \\ 0 & K_{\lambda_2} & \dots & \vdots & \vdots \\ \vdots & \dots & \dots & 0 & \vdots \\ 0 & \dots & 0 & K_{\lambda_p} & (C|D)^\perp \end{pmatrix} = (n+m)p$ , where  $\{\lambda_1, \lambda_2, \dots, \lambda_p\} = \sigma(A)$ ,

$$K_\lambda = \begin{pmatrix} M_\lambda & 0 \\ 0 & I_m \end{pmatrix} \text{ and } M_\lambda = (A_\lambda^{n_\lambda} | A_\lambda^{n_\lambda-1} B | \dots | A_\lambda B | B).$$

- (vi)  $\mathcal{G}_T^o := \int_0^T H_o(T, t)H_o(T, t)^T dt > 0$  for some  $T > 0$ , where  $H_o(T, t) = C \int_t^T e^{(T-\tau)A}Bd\tau + D$ .

## Theorem (Control computation)

Let  $(x_0, y_1) \in \mathbb{R}^n \times \mathbb{R}^q$ , and assume that system (1) is SOC.

For every  $T > 0$  and  $u_0 \in \mathbb{R}^m$ , the control steering  $x_0$  to  $y_1$  in time  $T$  is

$$u(t) = u_0 + \int_0^t H_o(T, \tau)^T d\tau (\mathcal{G}_T^o)^{-1} (y_1 - Ce^{TA}x_0 - H_o(T, 0)u_0) \quad (2)$$

Furthermore, this control is the unique minimizer of

$$\min \frac{1}{2} \int_0^T |\dot{u}(t)|_m^2 dt$$

|  $u \in H^1([0, T]; \mathbb{R}^m)$ ,  $u(0) = u_0$ ,  $y_1 = y_u(T, x_0)$ .

## Application to the cars example.

Consider for  $t \geq 0$ , system (1) with  $f_\ell = -\alpha_\ell v_\ell$  where  $v_\ell$  is the speed of car  $\ell$  for  $\ell = 1, 2$  and  $m_\ell$  stands for the mass of car  $\ell$ . Choosing  $m_1 = m_2 = 1$  and  $\alpha_1 = \alpha_2 = 1/2$  the system is given by

$$A = \begin{pmatrix} 0 & 1 & 0 & 0 \\ 0 & -\frac{1}{2} & 0 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & -\frac{1}{2} \end{pmatrix}, \quad B = \begin{pmatrix} 0 & 0 \\ 1 & 0 \\ 0 & 0 \\ 0 & 1 \end{pmatrix}, \quad C = \begin{pmatrix} 1 & 0 & -1 & 0 \\ 0 & 1 & 0 & -1 \\ 0 & -\frac{1}{2} & 0 & \frac{1}{2} \end{pmatrix} \text{ and } D = \begin{pmatrix} 0 & 0 \\ 0 & 0 \\ 1 & -1 \end{pmatrix}.$$

- One can see that this system is state to output controllable in any time  $T > 0$ .

- Goal: steer the system from  $x_0 = (1, 0, 1, 0)^T$  to  $y_1 = (2, 0, 0)^T$  in time  $T = 1$ .

Applying (2) with  $u_0 = (1, 0)^T$ , we get

$$u = (u_1, u_2)^T, \quad u_1 = 1 - u_2, \quad u_2(t) = at + bt^2 + c(e^{\frac{t}{2}} - 1) \text{ where } a = (254 + 98e - 316e^{\frac{1}{2}})/d,$$

$$b = (105 + 45e - 138e^{\frac{1}{2}})/d, \text{ and } c = (54 - 32e^{\frac{1}{2}})/d, \text{ with } d = 222e - 736e^{\frac{1}{2}} + 610$$

Using the matrix  $\mathcal{K}_T$ , we get  $u = (u_1, u_2)^T$ ,  $u_2 = -u_1$  and  $u_1(t) = (2e^{\frac{t}{2}} - e^{\frac{1}{2}} - 1)/(6e^{\frac{1}{2}} - 10)$  for  $t \in [0, 1)$  and  $u(1) = 0$ .

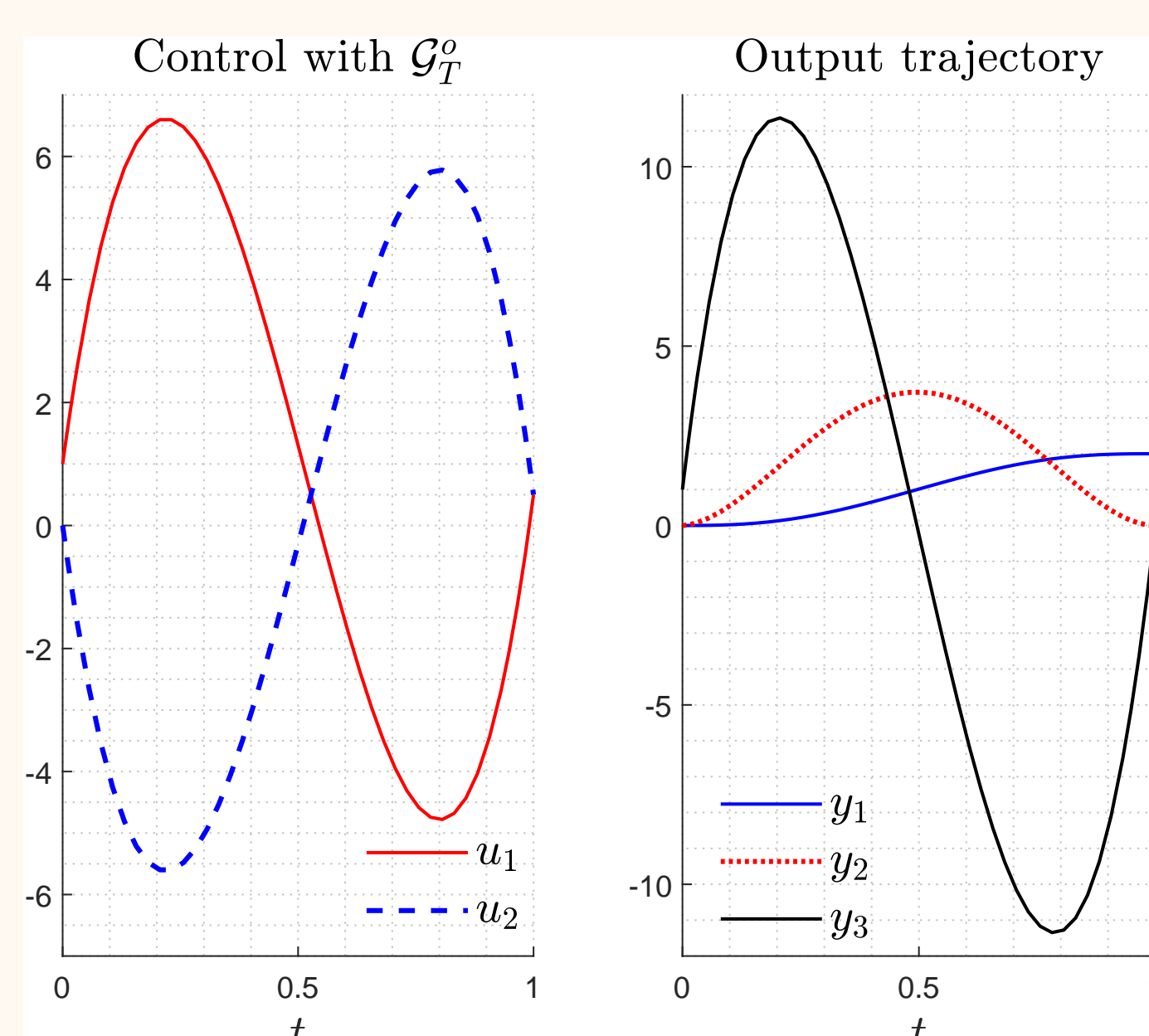


Figure: Continuous output trajectories with our matrix  $\mathcal{G}_T^o$ .

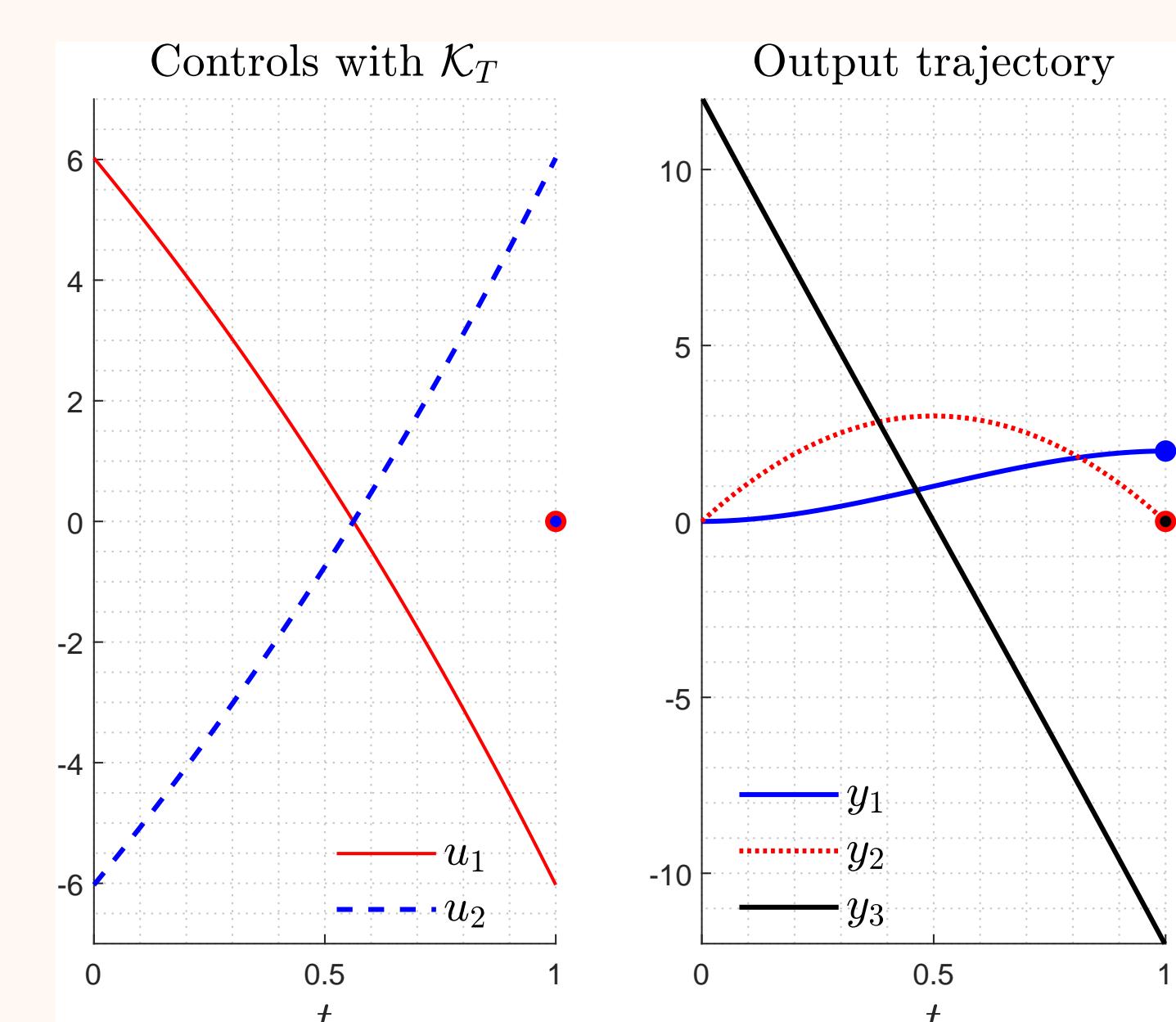


Figure: Discontinuous output trajectories with the matrix  $\mathcal{K}_T$  proposed by Kreindler and Sarachik.

## References

- J. Bertram and P. Sarachik. "On optimal computer control". *IFAC Proceedings Volumes 1.1* (1960).
- B. Danhane, J. Lohéac, and M. Jungers. "Characterizations of output controllability for LTI systems". *submitted, hal.03083128* (2020).
- E. Kreindler and P. Sarachik. "On the concepts of controllability and observability of linear systems". *IEEE Transactions on Automatic Control 9.2* (1964).